

# Bi-Directional GaNFast™

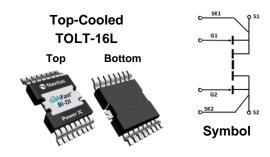


# 1. Features

- Bi-Directional 4-Quadrant GaN power switch:
  - Vss 650V continuous / 800V transient
- 52m  $\Omega$  Rss(ON)\_TYP\_25c and 49A Iss(CONT\_25C)
- TOLT-16L thermally-enhanced, top-cooled
- Zero reverse-recovery charge
- Up to 2MHz operation
- GaNFast<sup>™</sup> technology:
  - Integrated substrate clamp circuit between each source and the common substrate, optimizing switching performance in bi-directional current flow
- RoHS, Pb-free, REACH-compliant

# 2. Applications / Topologies

- Solar Micro-inverter / ESS and Heric Inverter
- Bi-Directional ZVS Cyclo-converter topologies
- AC-AC Motor Drive and Matrix Converters
- Next-generation Bi-Directional topologies

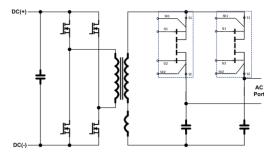


# 3. Description

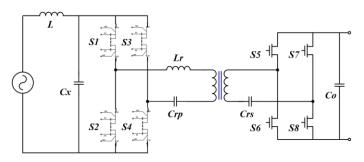
NV6428 is an **optimized** bi-directional switch capable of blocking voltage in both directions. A monolithic, integrated substrate clamping circuit between each source and the common substrate automatically clamps source-to-substrate voltage. Navitas' unique substrate clamp technology allows optimized switching performance during 4quadrant operation versus a **floating substrate** switch which can suffer 'back-gating effect'.

NV6428 also implements a thermally-enhanced top-cooled SMD with gull wing leads for superior board level temp cycling.

NV6428 is the ideal choice for topologies utilizing 4-quadrant switches to capture the benefits of bidirectional GaN for high-frequency, high-powerdensity, high-efficiency systems in solar, industrial, motor drive, and EV segments.



Cycloconverter



**Ultra-High Power Density Onboard Charger** 

# 4. Typical Application Circuits

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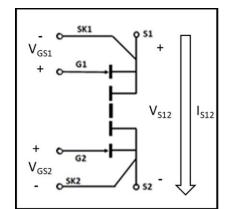
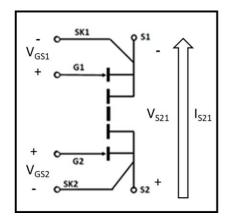


Fig. 1.  $V_{GS1}$ ,  $V_{GS2}$ ,  $V_{S12}$ ,  $I_{S12}$  Definition (Note 1 - 3)



**NV64** 

DATASHEE

Fig. 2.  $V_{GS1}$ ,  $V_{GS2}$ ,  $V_{S21}$ ,  $I_{S21}$  Definition (Note 1 - 3)

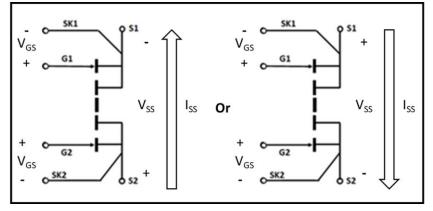


Fig. 3. V<sub>GS</sub>, V<sub>SS</sub>, I<sub>SS</sub> Definition (Note 4 - 6)

- (1)  $V_{GS1}$  implies that voltage is measured from gate\_1 to source\_1.
- (2)  $V_{S12}$  implies that voltage is measured from source\_1 to source\_2.
- (3)  $I_{S12}$  implies that current is flowing from source\_1 to source\_2.
- (4)  $V_{GS}$  implies that  $V_{GS} = V_{GS1} = V_{GS2}$ .
- (5) V<sub>SS</sub> implies that the source-to-source voltage can be applied in either direction.
- (6)  $I_{SS}$  implies that the source-to-source current can flow in either direction.
- (7) The device is symmetric. All measurements listed with respect to gate\_1 or source\_1 are the same when measured with respect to gate\_2 or source\_2.

# 7. Absolute Maximum Ratings (Note 8) (with respect to source, T<sub>CASE</sub> = 25°C, unless specified)

Symbol	Parameter	Max	Units
V <sub>SS_(CONT)</sub>	Continuous Source-to-Source Voltage	-650 to +650	V
$V_{SS_{TRAN}}$	Transient Source-to-Source Voltage (Note 9)	-800 to +800	V
V <sub>GS</sub>	Continuous Gate-to-Source Voltage	-10 to +7	V
$V_{\text{GS}\_(\text{TRAN})}$	Transient Gate-to-Source Voltage	-20 to +10	V
I <sub>SS_(CONT)</sub>	Continuous Current (T <sub>CASE</sub> = 25°C) Continuous Current (T <sub>CASE</sub> = 100°C, T <sub>JUNC</sub> = 150°C)	49 31	А
I <sub>SS_(PULSE)</sub>	Pulsed Current (10μs @ T <sub>JUNC</sub> = 25°C) Pulsed Current (10μs @ T <sub>JUNC</sub> = 150°C)	78 33	А
dV/dt	Source-to-Source Slew Rate	30	V/ns
T <sub>JUNC</sub>	Operating Junction Temperature	-40 to +150	°C
T <sub>STOR</sub>	Storage Temperature	-55 to +150	°C

(8) Absolute maximum ratings are stress ratings, and subjecting devices to stresses beyond these ratings may cause permanent damage.

(9)  $V_{SS (TRAN)}$  allows for surge ratings during *non-repetitive* events that are < 100 µs.

# 8. Recommended Operating Conditions (Note 10)

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Symbol	Parameter	Min	Тур	Max	Units
V <sub>GS</sub>	Gate Drive Voltage			6.5	V

(10) Exposure to conditions beyond maximum recommended operating conditions for extended periods of time may affect device reliability.

### 9. ESD Ratings

Symbol	Parameter	Мах	Units
CDM	Charged Device Model (per JS-002-2014)	750	V

### **10. Thermal Resistance**

Symbol	Parameter		Units
R <sub>e_JUNC-CASE</sub>	Junction-to-Case Thermal Resistance		°C/W

# **11. Electrical Characteristics**

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Conditions unless specified:  $V_{co} = 400V$ ,  $V_{co} = 6.5V$ ,  $T_{coor} = 25^{\circ}C$ ,  $I_{co} = 12A$ 

Symbol	unless specified: V <sub>ss</sub> = 400V, <b>Parameter</b>	Min	Тур	Max	Units	Conditions
	t GaN Switch Characteristic	S		ı	• 	1
I <sub>SSS</sub>	Source-Source Leakage Current		80		μA	V <sub>SS</sub> = 650V, V <sub>GS</sub> = 0V
I <sub>SSS</sub>	Source-Source Leakage Current		110		μA	$V_{SS} = 650V, V_{GS} = 0V,$ $T_{JUNC} = 150^{\circ}C$
I <sub>GSS</sub>	Gate-Source Leakage Current		160		μA	V <sub>GS</sub> = 6.5V
R <sub>SS(ON)</sub>	Source-Source Resistance		52	70	mΩ	V <sub>GS</sub> = 6.5V, I <sub>SS</sub> = 12A
R <sub>SS(ON)</sub>	Source-Source Resistance		125		mΩ	$V_{GS} = 6.5V$ , $I_{SS} = 12A$ , $T_{JUNC} = 150^{\circ}C$ (by Design)
$V_{GS1(th)}$	Gate Threshold Voltage	1	1.5	2.8	V	$I_{S21} = 23mA, V_{S21} = 0.1V$
$V_{S12\_Reverse}$	Source-Source Third Quadrant Conduction Voltage		3.3	5	V	$V_{GS1} = 0V, V_{GS2} = 6.5V,$ $I_{S12} = 12A$
Q <sub>RR</sub>	Reverse Recovery Charge		Zero		nC	
$R_{G}$	Internal Gate Resistance		400		mΩ	By Design
C <sub>ISS</sub>	Input Capacitance		254		pF	V <sub>SS</sub> = 400V, V <sub>GS</sub> = 0V
C <sub>ISS1</sub>	Input Capacitance		254		pF	$V_{S21} = 400V, V_{GS1} = 0V, V_{GS2} = 6.5V$
C <sub>oss</sub>	Output Capacitance		145		pF	V <sub>SS</sub> = 400V, V <sub>GS</sub> = 0V
C <sub>OSS1</sub>	Output Capacitance		155		pF	$V_{S21} = 400V, V_{GS1} = 0V, V_{GS2} = 6.5V$
Crss	Reverse Transfer Capacitance		2		pF	V <sub>SS</sub> = 400V, V <sub>GS</sub> = 0V
Crss1	Reverse Transfer Capacitance		2		pF	$V_{S21} = 400V, V_{GS1} = 0V,$ $V_{GS2} = 6.5V$
$Q_{\mathrm{G}}$	Total Gate Charge for Each Gate		6.49		nC	$V_{S21} = 400V, V_{GS1} = 0V \text{ to } 6.5V, V_{GS2} = 0V$
Q <sub>G1</sub>	Total Gate Charge for Each Gate		7.51		nC	$V_{S21} = 400V, V_{GS1} = 0V \text{ to } 6.5V, V_{GS2} = 6.5V$
Q <sub>GS</sub>	Gate-Source Charge for Each Individual Gate		1.51		nC	$V_{S21} = 400V, V_{GS1} = 0V \text{ to } 6.5V, V_{GS2} = 0V$
Q <sub>GS1</sub>	Gate-Source Charge for Each Individual Gate		1.51		nC	$V_{S21} = 400V, V_{GS1} = 0V \text{ to } 6.5V, V_{GS2} = 6.5V$

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Q <sub>oss</sub>	Output Charge	80	nC	V <sub>SS</sub> = 400V, V <sub>GS</sub> = 0V
Q <sub>OSS1</sub>	Output Charge	100	nC	V <sub>S21</sub> = 400V, V <sub>GS1</sub> = 0V, V <sub>GS2</sub> = 6.5V
C <sub>O(er)</sub> (Note 10)	Effective Output Capacitance, Energy Related	170	pF	V <sub>SS</sub> = 400V, V <sub>GS</sub> = 0V
C <sub>O(er)1</sub> (Note 11)	Effective Output Capacitance, Energy Related	190	pF	V <sub>S21</sub> = 400V, V <sub>GS1</sub> = 0V, V <sub>GS2</sub> = 6.5V
C <sub>O(tr)</sub> (Note 12)	Effective Output Capacitance, Time Related	195	pF	V <sub>SS</sub> = 400V, V <sub>GS</sub> = 0V
CO(tr)1 (Note 13)	Effective Output Capacitance, Time Related	250	pF	$V_{S21} = 400V, V_{GS1} = 0V,$ $V_{GS2} = 6.5V$

(10)  $C_{O(er)}$  is a fixed capacitance that gives the same stored energy as  $C_{OSS}$  while  $V_{SS}$  rises from 0V to 400V.

(11)  $C_{O(er)1}$  is a fixed capacitance that gives the same stored energy as  $C_{OSS1}$  while  $V_{SS}$  rises from 0V to 400V.

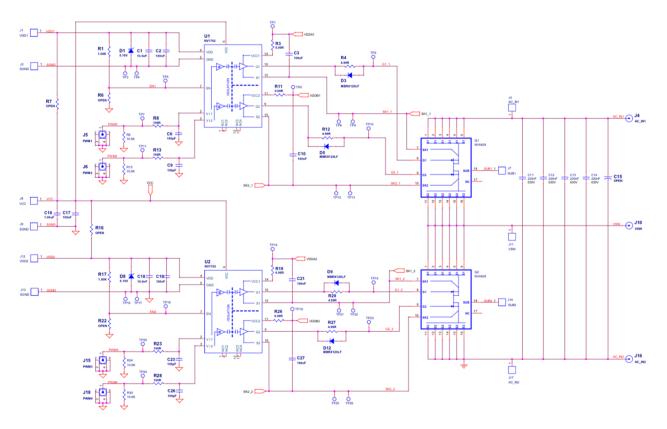
(12)  $C_{O(tr)}$  is a fixed capacitance that gives the same charging time as  $C_{OSS}$  while  $V_{SS}$  rises from 0V to 400V.

(13)  $C_{O(tr)1}$  is a fixed capacitance that gives the same charging time as  $C_{OSS1}$  while  $V_{SS}$  rises from 0V to 400V.



# 12. Inductive Switching Test Circuit

### Schematic 1. Inductive Switching Test Circuit





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# 13. Electrical Curves (GaN FET, T<sub>CASE</sub> = 25°C unless otherwise specified)

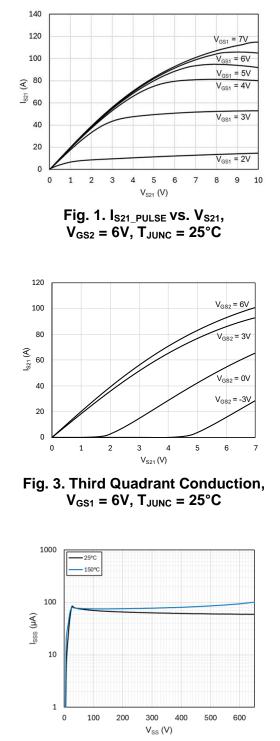
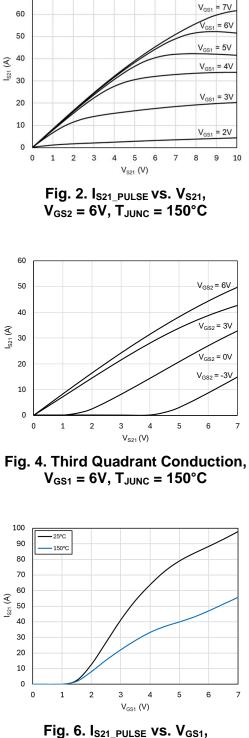


Fig. 5. I<sub>SSS</sub> vs. V<sub>SS</sub>, T<sub>JUNC</sub> = 25°C, 150°C

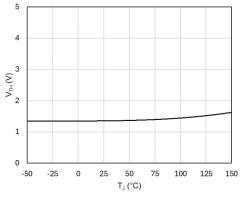


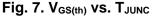
 $V_{GS2} = 6V, V_{S21} = 10V$ 

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# Bi-Directional GaNFast NV6428

#### DATASHEET





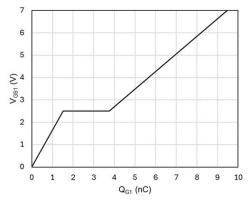


Fig. 9.  $V_{GS1}$  vs.  $Q_{G1}$ ,  $V_{GS2} = 6V$ 

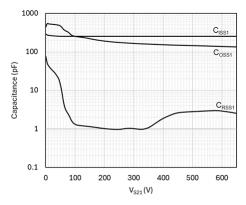


Fig. 11.  $C_{ISS1}$ ,  $C_{OSS1}$ ,  $C_{RSS1}$  vs.  $V_{S21}$ ,  $V_{GS1} = 0V, V_{GS2} = 6V$ 

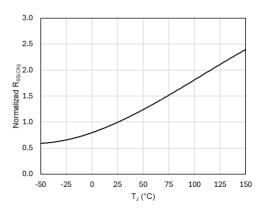


Fig. 8. Normalized RSS(ON) vs. TJUNC

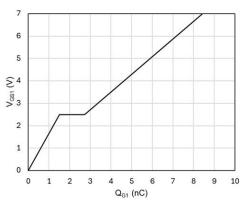


Fig. 10.  $V_{GS1}$  vs.  $Q_{G1}$ ,  $V_{GS2} = 0V$ 

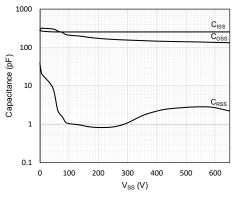
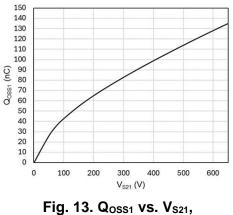


Fig. 12.  $C_{ISS}$ ,  $C_{OSS}$ ,  $C_{RSS}$  vs.  $V_{SS}$ ,  $V_{GS} = 0V$ 

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# Bi-Directional GaNFast NV6428

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 $V_{GS1} = 0V, V_{GS2} = 6V$ 

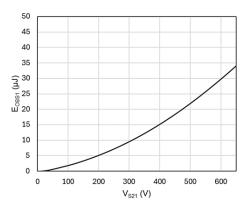


Fig. 15. E<sub>OSS1</sub> vs. V<sub>S21</sub>,  $V_{GS1} = 0V, V_{GS2} = 6V$ 

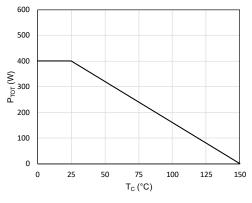


Fig. 17. PDISSIPATION VS. TCASE

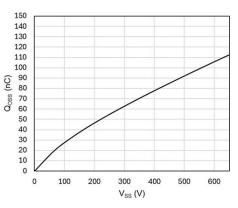


Fig. 14.  $Q_{OSS}$  vs.  $V_{SS}$ ,  $V_{GS}$  = 0V

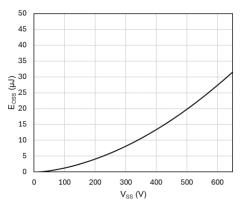


Fig. 16.  $E_{OSS}$  vs.  $V_{SS}$ ,  $V_{GS} = 0V$ 

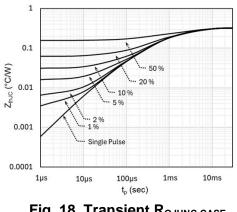


Fig. 18. Transient ROJUNC-CASE



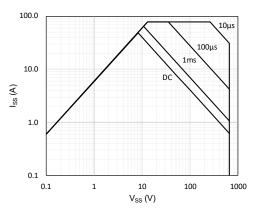
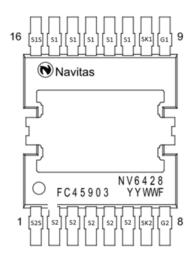


Fig. 19. Safe Operating Area, T<sub>JUNC</sub> = 25°C

DATASHEET

# 14. Pinout Table



Pi	n	1/0	Description	
Number	Symbol	1/0	Description	
1	S2S	Р	Connect to S2 on PCB	
2-6	S2	Р	Source 2 Terminal	
7	SK2	G	Kelvin Source 2	
8	G2	Ι	Gate 2	
9	G1	Ι	Gate 1	
10	SK1	G	Kelvin Source 1	
11-15	S1	Р	Source 1 Terminal	
16	S1S	Р	Connect to S1 on PCB	
Top Pad	N/A	N/A	Substrate: Requires Isolation to Heatsink	
	Note: I = Input, P = Power, G = Ground			

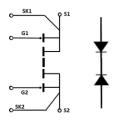


## **15. Functional Description**

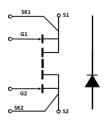
### **15.1. Operating Modes**

NV6428 is a normally-off GaNFET device with two gates (G1, G2) and two sources (S1, S2). G1 acts with respect to S1, and G2 acts with respect to S2, but G2 does not act with respect to S1, and vice-versa. As a 4-Quadrant Switch (4QS), it is capable of blocking voltage in either or both directions, and conducting current in either or both directions, depending on the states of G1 and G2.

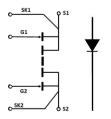
**Mode 1** ( $V_{GS1} = 0V$ ,  $V_{GS2} = 0V$ ): Voltage is blocked in both the  $V_{S12}$  and  $V_{S21}$  direction. Current is blocked in both the  $I_{S12}$  and  $I_{S21}$  direction.



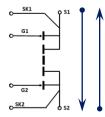
**Mode 2** ( $V_{GS1} = 6.5V$ ,  $V_{GS2} = 0V$ ): Voltage is blocked in the  $V_{S12}$  direction only. Current flows in the  $I_{S21}$  direction but is blocked in the  $I_{S12}$  direction.  $V_{S21}$  should not exceed 7V.



**Mode 3** ( $V_{GS1} = 0V$ ,  $V_{GS2} = 6.5V$ ): Voltage is blocked in the  $V_{S21}$  direction only. Current flows in the  $I_{S12}$  direction but is blocked in the  $I_{S21}$  direction.  $V_{S12}$  should not exceed 7V.



**Mode 4** ( $V_{GS1} = 6.5V$ ,  $V_{GS2} = 6.5V$ ): Current can flow in either direction. If  $V_{S12} > V_{S21}$ , then current flows in the  $I_{S12}$  direction. If  $V_{S21} > V_{S12}$ , then current flows in the  $I_{S21}$  direction.





#### 15.2. Integrated Substrate Clamp

NV6428 has a monolithic, integrated substrate clamping circuit which optimizes the silicon substrate potential under any bias condition applied between the two sources. This clamp circuit prevents undesired  $R_{SS(ON)}$  increases that can happen when the silicon substrate potential is uncontrolled.

This major performance enhancement is possible by virtue of Navitas *GaNFast*<sup>™</sup> technology and this functionality differentiates NV6428 from competitor parts.

### 15.3. Gate Driver and Gate Power Supply Selection

NV6428 Bi-Directional Switch requires two floating gate driver channels to provide PWM-generated gate drive inputs to the two gates. An isolated driver is recommended, providing 6V~6.5V output voltage for best performance and should be able to turn on both gate outputs simultaneously (Mode 4 operation). The type of isolation is dependent on the circuit and system requirements. Functional isolation may be enough in many cases, but in some cases safety isolation may also be needed.

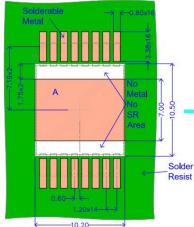
At least one of the two gates will not be referenced to ground or a DC rail and will require a drive channel that can float to high voltage that matches the range of the source it is driving. In some applications, both sources are switching nodes, so both gate drive outputs and the power supplies for them must be able to float to remain referenced to the source being driven. In some circuits, a ground referenced power supply and a bootstrap power supply is sufficient, while in others an isolated floating power supply for one or both gate drives may be required.

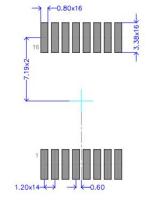


### 15.4. PCB Layout Guidelines and PCBA SMT Guidelines

PCB layout is critical for thermal management, noise immunity, and proper operation of the device. The following rules should be followed carefully during the design of the PCB layout:

• Do not run power SOURCE current through SK pin!





### PCBA SMT Guidelines:

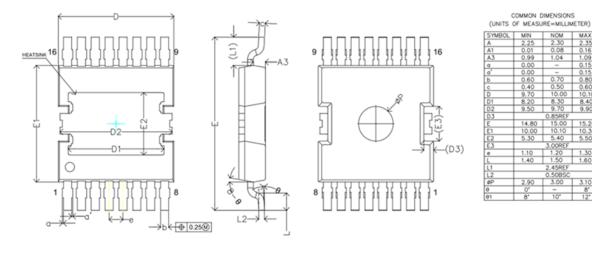
- 1.6mm thick FR4, 4-Layer 2-Oz Cu
- Solder Mask per DWG on left
- Solder Stencil per Customer's PCBA SMT
- Solder Reflow Profile per PCBA SMT Vendor
- Recommended Solder Paste: SAC305
- Recommended 10% maximum Voids

RECOMMENDED PCB FOOTPRINT

RECOMMENDED STENCIL OPENING



# 16. Package Outline Dimensions:



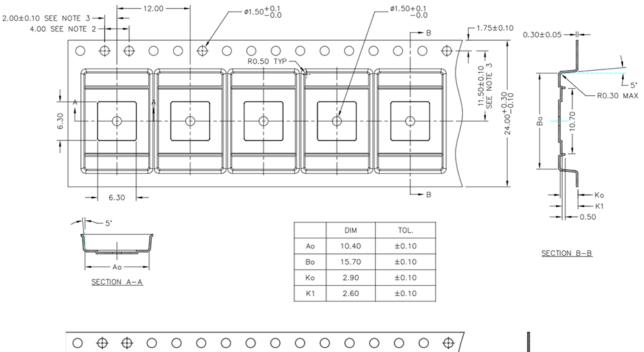


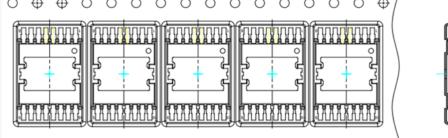


NOTES: 1.ALL DIMENSIONS DO NOT INCLUDE MOLD FLASH OR PROTRUSION.



# 17. TnR Drawing





# 18. Ordering Information

Part Number	Qualification	Package	MSL Rating	TnR Dia. and Qty
NV6428	JEDEC	TOLT-16L Top-cooled SMD	2	Standard (13" dia) Qty1,500
NV6428-RA	JEDEC		3	Mini-Reel (7" dia) Qty450

# 19. Revision History

Date	Status	Notes
Apr 2 <sup>nd</sup> , 2024	Revision	Updated Electrical Characteristics and Curves data
Oct 11 <sup>th</sup> , 2024	Revision	Updated Electrical Characteristics and Curves data
Jan 23 <sup>rd</sup> , 2025	Revision	Updated Inductive Switching Reference Schematic and Electrical Characteristics
Mar 6 <sup>th</sup> , 2025	Final	Updated Features section





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